

# Applications of free probability and random matrix theory

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# Some important concepts from classical probability

- ▶ Random variables are functions (i.e. they commute w.r.t. multiplication) with a given p.d.f. (denoted  $f$ )
- ▶ Expectation (denoted  $E$ ) is integration
- ▶ Independence
- ▶ Additive convolution ( $*$ ) and the logarithm of the Fourier transform
- ▶ Multiplicative convolution
- ▶ Central limit law, with special role of the Gaussian law
- ▶ Poisson distribution  $P_c$ : The limit of  $\left(\left(1 - \frac{c}{n}\right) \delta(0) + \frac{c}{n} \delta(1)\right)^{*n}$  as  $n \rightarrow \infty$ .
- ▶ Divisibility: For a given  $a$ , find i.i.d.  $b_1, \dots, b_n$  such that  $f_a = f_{b_1 + \dots + b_n}$ .

- ▶ Can we find a more general theory, where the random variables are matrices (or more generally, operators), with their eigenvalue distribution (or spectrum) taking the role as the p.d.f.?
- ▶ What are the analogues to the above mentioned concepts for this theory? In particular, what plays the role as central limit?
- ▶ What are the applications of such a theory?

# Free probability

Free probability was developed as a probability theory for random variables which do not commute, like matrices

- ▶ The random variables are elements in a unital  $*$ -algebra (denoted  $A$ ), typically  $B(\mathcal{H})$ , or  $M_n(\mathbb{C})$ .
- ▶ Expectation (denoted  $\phi$ ) is a normalized linear functional on  $A$ . The pair  $(A, \phi)$  is called a *noncommutative probability space*.
- ▶ For matrices,  $\phi$  will be the normalized trace  $tr_n$ , defined by

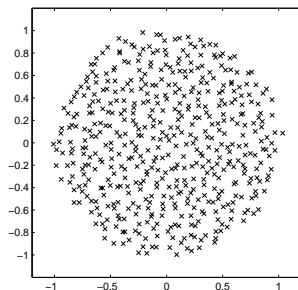
$$tr_n(a) = \frac{1}{n} \sum_{i=1}^n a_{ii}.$$

For random matrices, we set  $\phi(a) = \tau_n(a) = E(tr_n(a))$  is defined by.

- ▶ Useful way to think about free probability: Two independent random matrices, where the eigenvectors for one of them point in each direction with equal probability, then the two matrices are "free" (to be defined).

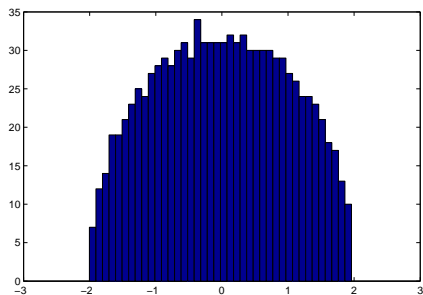
# The full circle law

Let  $\mathbf{X}_n = \frac{1}{\sqrt{n}}\mathbf{Y}_n$  where  $\mathbf{Y}_n$  is  $n \times n$  and has i.i.d. complex standard Gaussian entries. This is a "central limit candidate". What happens when  $n$  is large? The eigenvalues converge to what is called the *full circle law*. Here for  $n = 500$ .



```
plot(eig( (1/sqrt(1000)) * (randn(500,500) +  
j*randn(500,500)) ), 'kx')
```

# The semicircle law



```
A = (1/sqrt(2000)) * (randn(1000,1000) +  
j*randn(1000,1000));  
A = (sqrt(2)/2)*(A+A');  
hist(eig(A),40)
```

# The Marčenko Pastur law

What happens with the eigenvalues of  $\frac{1}{N}\mathbf{X}_n\mathbf{X}_n^H$  when  $\mathbf{X}_n$  is an  $n \times N$  random matrix with standard complex Gaussian entries?

- ▶ One can show that

$$\tau_n \left( \left( \frac{1}{N} \mathbf{X}_n \mathbf{X}_n^H \right)^p \right) = \sum_{\hat{\pi} \in NC_{2p}} 1 + \sum_k \frac{a_k}{N^{2k}}.$$

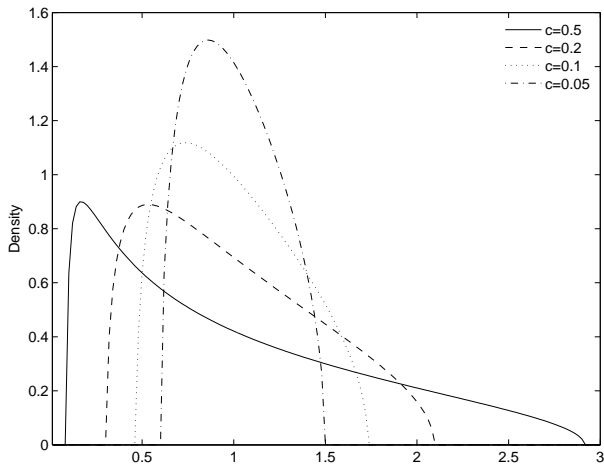
Convergence is "almost sure", i.e. we have very accurate eigenvalue prediction when the matrices are large.

- ▶ When  $\frac{n}{N} \rightarrow c$ , the eigenvalue distribution converges to the Marčenko Pastur law with parameter  $c$ , denoted  $\mu_{\frac{n}{N}}$ , with density

$$f^{\mu_c}(x) = \left(1 - \frac{1}{c}\right)^+ \delta(x) + \frac{\sqrt{(x-a)^+(b-x)^+}}{2\pi cx}, \quad (1)$$

where  $(z)^+ = \max(0, z)$ ,  $a = (1 - \sqrt{c})^2$  and  $b = (1 + \sqrt{c})^2$ .

Four different Marčenko Pastur laws  $\mu_{\frac{n}{N}}$  are drawn.





## Motivation for free probability

One can show that for the Gaussian random matrices we considered, the limits

$$\phi(X^{i_1} Y^{j_1} \dots X^{i_l} Y^{j_l}) = \lim_{n \rightarrow \infty} \text{tr}_n(\mathbf{X}_n^{i_1} \mathbf{Y}_n^{j_1} \dots \mathbf{X}_n^{i_l} \mathbf{Y}_n^{j_l})$$

exist. If we linearly extend the linear functional  $\phi$  to all polynomials in  $A$  and  $B$ , the following can be shown:

### Theorem

*If  $P_i, Q_i$  are polynomials in  $X$  and  $Y$  respectively, with  $1 \leq i \leq l$ , and  $\phi(P_i(X)) = 0, \phi(Q_i(Y)) = 0$  for all  $i$ , then*

$$\phi(P_1(X)Q_1(Y) \dots P_l(X)Q_l(Y)) = 0.$$

This motivates the definition of freeness, which is the analogy to independence.

# Definition of freeness

## Definition

A family of unital  $*$ -subalgebras  $(A_i)_{i \in I}$  is called a free family if

$$\left\{ \begin{array}{l} a_j \in A_{i_j} \\ i_1 \neq i_2, i_2 \neq i_3, \dots, i_{n-1} \neq i_n \\ \phi(a_1) = \phi(a_2) = \dots = \phi(a_n) = 0 \end{array} \right\} \Rightarrow \phi(a_1 \cdots a_n) = 0. \quad (2)$$

A family of random variables  $a_i$  is called a free family if the algebras they generate form a free family.

# The free central limit theorem

## Theorem

If

- ▶  $a_1, \dots, a_n$  are free and self-adjoint,
- ▶  $\phi(a_i) = 0$ ,
- ▶  $\phi(a_i^2) = 1$ ,
- ▶  $\sup_i |\phi(a_i^k)| < \infty$  for all  $k$ ,

then the sequence  $(a_1 + \dots + a_n)/\sqrt{n}$  converges in distribution to the semicircle law.

In free probability, the semicircle law thus has the role of the Gaussian law. its density is density  $\frac{1}{2\pi} \sqrt{4 - x^2}$

# Similarities between classical and free probability

1. Additive convolution  $\boxplus$ : The p.d.f. of the sum of free random variables. The role of the logarithm of the Fourier transform is now taken by the  $R$ -transform, which satisfies
 
$$R_{\mu_a \boxplus \mu_b}(z) = R_{\mu_a}(z) + R_{\mu_b}(z).$$
2. The  $S$ -transform: Transform on probability distributions which satisfies
 
$$S_{\mu_a \boxtimes \mu_b}(z) = S_{\mu_a}(z)S_{\mu_b}(z)$$
3. Poisson distributions have their analogy in the free Poisson distributions: These are given by the Marčenko Pastur laws  $\mu_c$  with parameter  $c$ , which also can be written as the limit of
 
$$\left( \left(1 - \frac{c}{n}\right) \delta(0) + \frac{c}{n} \delta(1) \right)^{\boxplus n}$$
 as  $n \rightarrow \infty$
4. Infinite divisibility: There exists an analogy to the Lévy-Hinčin formula for infinite divisibility in classical probability.

# Main usage of free probability in my papers

- ▶ Let  $\mathbf{X}$  and  $\mathbf{Y}$  be random matrices. How can we make a good prediction of the eigenvalue distribution of  $\mathbf{X}$  when one has the eigenvalue distribution of  $\mathbf{XY}$  and  $\mathbf{Y}$ ? Simplest case is when one assumes that  $\mathbf{Y}$  is Gaussian (Noise). What about the eigenvectors?
- ▶ Assume that we have the eigenvalue distribution of  $\frac{1}{N}(\mathbf{R} + \mathbf{X})(\mathbf{R} + \mathbf{X})^H$ , where  $\mathbf{R}$  and  $\mathbf{X}$  are independent  $n \times N$  random matrices, with  $\mathbf{X}$  Gaussian. If the columns of  $\mathbf{R}$  are realizations of some random vector  $\mathbf{r}$ , what is the covariance matrix  $E(\mathbf{r}_i \mathbf{r}_j^*)$ ?
- ▶ Multiplicative free convolution with the Marčenko Pastur law is particularly useful. This has an efficient implementation.

# Application of free probability

The capacity per receiving antenna of a channel with  $n \times m$  channel matrix  $\mathbf{H}$  and signal to noise ratio  $\rho = \frac{1}{\sigma^2}$  is given by

$$C = \frac{1}{n} \log_2 \det \left( \mathbf{I}_n + \frac{1}{m\sigma^2} \mathbf{H}\mathbf{H}^H \right) = \frac{1}{n} \sum_{l=1}^n \log_2 \left( 1 + \frac{1}{\sigma^2} \lambda_l \right) \quad (3)$$

where  $\lambda_l$  are the eigenvalues of  $\frac{1}{m} \mathbf{H}\mathbf{H}^H$ . We would like to estimate  $C$ .

To estimate  $C$ , we will use free probability tools to estimate the eigenvalues of  $\frac{1}{m} \mathbf{H}\mathbf{H}^H$  based on some observations  $\hat{\mathbf{H}}$ ;

# Observation model

The following is a much used observation model:

$$\hat{\mathbf{H}}_i = \mathbf{H} + \sigma \mathbf{X}_i \quad (4)$$

where

- ▶ The matrices are  $n \times m$  ( $n$  is the number of receiving antennas,  $m$  is the number of transmitting antennas)
- ▶  $\hat{\mathbf{H}}_i$  is the measured MIMO matrix,
- ▶  $\mathbf{X}_i$  is the noise matrix with i.i.d standard complex Gaussian entries.

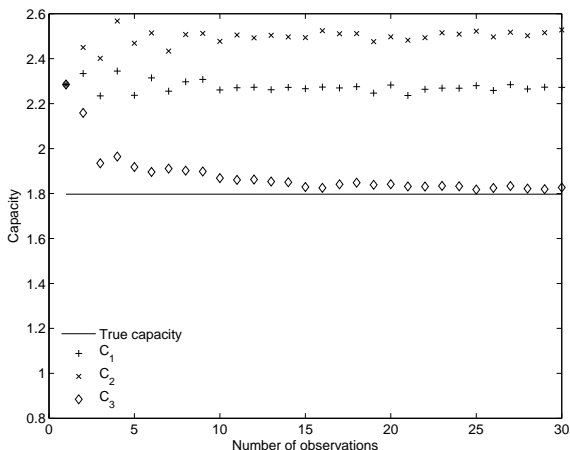
# Existing ways to estimate the channel capacity

Several channel capacity estimators have been used in the literature:

$$\begin{aligned}
 C_1 &= \frac{1}{nL} \sum_{i=1}^L \log_2 \det \left( \mathbf{I}_n + \frac{1}{m\sigma^2} \hat{\mathbf{H}}_i \hat{\mathbf{H}}_i^H \right) \\
 C_2 &= \frac{1}{n} \log_2 \det \left( \mathbf{I}_n + \frac{1}{L\sigma^2 m} \sum_{i=1}^L \hat{\mathbf{H}}_i \hat{\mathbf{H}}_i^H \right) \\
 C_3 &= \frac{1}{n} \log_2 \det \left( \mathbf{I}_n + \frac{1}{\sigma^2 m} \left( \frac{1}{L} \sum_{i=1}^L \hat{\mathbf{H}}_i \right) \left( \frac{1}{L} \sum_{i=1}^L \hat{\mathbf{H}}_i \right)^H \right)
 \end{aligned} \tag{5}$$

Why not try to formulate an estimator based on free probability instead?





Comparison of the classical capacity estimators for various number of observations.  $\sigma^2 = 0.1$ ,  $n = 10$  receive antennas,  $m = 10$  transmit antennas. The rank of  $\mathbf{H}$  was 3.

# Main free probability result we will use

Define

$$\begin{aligned}\Gamma_n &= \frac{1}{N} \mathbf{R}_n \mathbf{R}_n^H \\ \mathbf{W}_n &= \frac{1}{N} (\mathbf{R}_n + \sigma \mathbf{X}_n) (\mathbf{R}_n + \sigma \mathbf{X}_n)^H,\end{aligned}$$

where  $\mathbf{R}_n$  and  $\mathbf{X}_n$  are independent  $n \times N$  random matrices,  $\mathbf{X}_n$  is complex, standard, Gaussian.

## Theorem

If  $e.e.d.(\Gamma_n) \rightarrow \nu_\Gamma$ , then  $e.e.d.(\mathbf{W}_n) \rightarrow \nu_W$  where  $\nu_W$  is uniquely identified by

$$\nu_W \boxminus \mu_c = (\nu_\Gamma \boxminus \mu_c) \boxplus \delta_{\sigma^2}$$

( $\boxminus$  = "the opposite of  $\boxplus$ ").

# Realization of the theorem for the problem at hand

Form the compound observation matrix

$$\hat{\mathbf{H}}_{1\dots L} = \mathbf{H}_{1\dots L} + \frac{\sigma}{\sqrt{L}} \mathbf{X}_{1\dots L}, \text{ where}$$

$$\hat{\mathbf{H}}_{1\dots L} = \frac{1}{\sqrt{L}} [\hat{\mathbf{H}}_1, \hat{\mathbf{H}}_2, \dots, \hat{\mathbf{H}}_L],$$

$$\mathbf{H}_{1\dots L} = \frac{1}{\sqrt{L}} [\mathbf{H}, \mathbf{H}, \dots, \mathbf{H}],$$

$$\mathbf{X}_{1\dots L} = [\mathbf{X}_1, \mathbf{X}_2, \dots, \mathbf{X}_L].$$

For the problem at hand, the theorem takes the form

$$\nu_{\frac{1}{m} \hat{\mathbf{H}}_{1\dots L} \hat{\mathbf{H}}_{1\dots L}^H} \boxtimes \mu_{\frac{n}{mL}} \approx \left( \nu_{\frac{1}{m} \mathbf{H}_{1\dots L} \mathbf{H}_{1\dots L}^H} \boxtimes \mu_{\frac{n}{mL}} \right) \boxplus \delta_{\sigma^2} \quad (6)$$

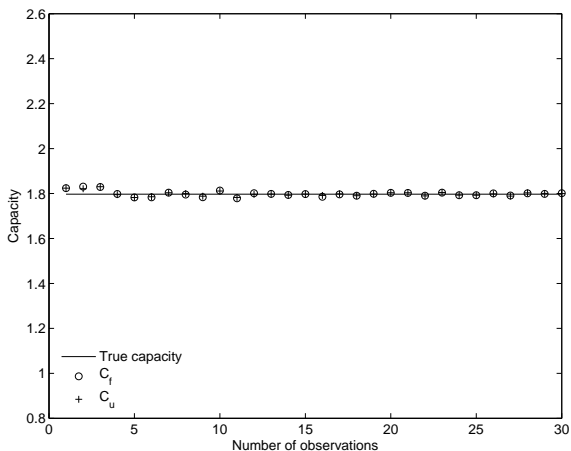
Since  $\frac{1}{m} \mathbf{H}_{1\dots L} \mathbf{H}_{1\dots L}^H = \frac{1}{m} \mathbf{H} \mathbf{H}^H$ , we can now estimate the moments of  $\frac{1}{m} \mathbf{H} \mathbf{H}^H$  from the moments of the observation matrix  $\frac{1}{m} \hat{\mathbf{H}}_{1\dots L} \hat{\mathbf{H}}_{1\dots L}^H$ , and thereby estimate the eigenvalues, and hence the channel capacity.

# Free probability based estimator for the moments of the channel matrix

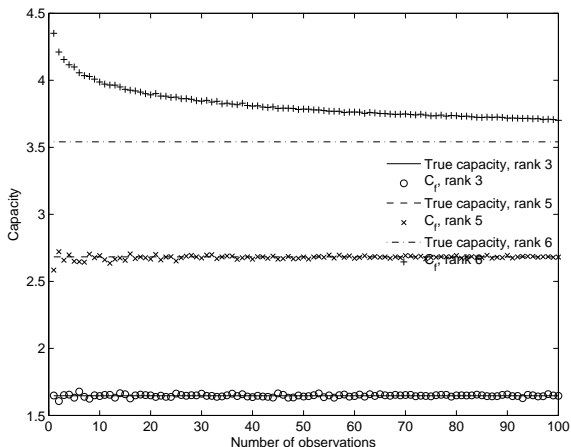
Can also be written in the following way for the first four moments:

$$\begin{aligned}
 \hat{h}_1 &= h_1 + \sigma^2 \\
 \hat{h}_2 &= h_2 + 2\sigma^2(1+c)h_1 + \sigma^4(1+c) \\
 \hat{h}_3 &= h_3 + 3\sigma^2(1+c)h_2 + 3\sigma^2ch_1^2 \\
 &\quad + 3\sigma^4(c^2 + 3c + 1)h_1 \\
 &\quad + \sigma^6(c^2 + 3c + 1) \\
 \hat{h}_4 &= h_4 + 4\sigma^2(1+c)h_3 + 8\sigma^2ch_2h_1 \\
 &\quad + \sigma^4(6c^2 + 16c + 6)h_2 \\
 &\quad + 14\sigma^4c(1+c)h_1^2 \\
 &\quad + 4\sigma^6(c^3 + 6c^2 + 6c + 1)h_1 \\
 &\quad + \sigma^8(c^3 + 6c^2 + 6c + 1),
 \end{aligned} \tag{7}$$

where  $\hat{h}_i$  are the moments of the observation matrix  $\frac{1}{m}\hat{\mathbf{H}}_{1\dots L}\hat{\mathbf{H}}_{1\dots L}^H$ ,  
 $h_i$  are the moments of  $\frac{1}{m}\mathbf{H}\mathbf{H}^H$ .



Comparison of  $C_f$  and  $C_u$  for various number of observations.  
 $\sigma^2 = 0.1$ ,  $n = 10$  receive antennas,  $m = 10$  transmit antennas. The rank of  $\mathbf{H}$  was 3.



$C_f$  for various number of observations. No phase off-set/phase drift.  $\sigma^2 = 0.1$ ,  $n = 10$  receive antennas,  $m = 10$  transmit antennas. The rank of  $\mathbf{H}$  was 3, 5 and 6.

## List of papers

- ▶ Free Deconvolution for Signal Processing Applications. Submitted to IEEE Trans. Inform. Theory. [arxiv.org/cs.IT/0701025](http://arxiv.org/cs.IT/0701025).
- ▶ Multiplicative free Convolution and Information-Plus-Noise Type Matrices. [arxiv.org/math.PR/0702342](http://arxiv.org/math.PR/0702342).
- ▶ Channel Capacity Estimation using Free Probability Theory. Submitted to IEEE. Trans. Signal Process. [arxiv.org/abs/0707.3095](http://arxiv.org/abs/0707.3095).
- ▶ Random Vandermonde Matrices-Part I: Fundamental results. Work in progress.
- ▶ Random Vandermonde Matrices-Part II: Applications to wireless applications. Work in progress.
- ▶ Vandermonde Frequency Division Multiplexing: An Approach for Cognitive Radio. Work in progress.

This talk is available at

<http://heim.ifi.uio.no/~oyvindry/talks.shtml>.

My publications are listed at

<http://heim.ifi.uio.no/~oyvindry/publications.shtml>